

2012-11-12

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Recommended Citation

Ammann, M., Bao, X. L., & Narbudowicz, A. (2012) Circularly Polarized Terminal Antennas for Emerging Wireless Systems, LAPC - *Loughborough Antennas & Propagation Conference*, Loughborough, UK, pp. 1-5, 12/11/2012. doi:10.1109/LAPC.2012.6402946

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Circularly Polarized Terminal Antennas for Emerging Wireless Systems

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Abstract—Several types of omni-directional Circularly Polarized (CP) antennas are presented, which are employed by three different types of feedlines: Coplanar Waveguide (CPW) fed, Microstrip fed, and Differential-fed, to achieve omni-directional CP performances. The Dual-frequency Omni-directional CP antenna is proposed by using slot and inductor embedded into radiated patches. These omni-directional CP antennas have potential application on GPS, WLAN and RFID systems.

INTRODUCTION

With the increased demand for wireless network systems, satellite communications systems and global navigation satellite systems, circularly polarized antennas have become more attractive to engineering designers [1-3]. Circular polarization (CP) combats Faraday rotation and can significantly reduce multipath fading effects and polarization mismatch loss experienced in linear systems. Many techniques have been reported to improve the performance of CP antennas, in particular with improvements in bandwidth, gain and beamwidth. The antenna bandwidth and gain of a rhombic loop antenna were increased by using coupling strips [4]. A combined structure with two bowtie patches and two electric dipoles were used to provide wide beamwidth CP [5]. A reconfigurable patch antenna was introduced to obtain dual-sense CP characteristics [6].

In recent years, omni-directional CP antennas have been presented in the literature using various techniques [7-11]. One method is to use back-to-back microstrip patches which are fed by coplanar waveguide (CPW) [7]. The combined radiation from the patches provide omni-directional CP performance in one plane. Another technique for low-profile application uses a top-loaded cylindrical monopole structure and four printed arc-shaped dipoles providing omni-directional CP in the azimuth plane [10]. Additionally, a conical skirt monopole antenna with a polarizer was presented [11]. The polarizer consists of a series of 45° tilted parasitic strips embedded into the surface of the cylinder dielectric. The polarizer can convert the wave from linear to circular polarization. Hence, omni-directional CP performance is achieved. More recently, a zero-order resonator (ZOR) loaded

with arc-shaped subs along the patch was employed to achieve CP omni-directional performance [12, 13].

In this paper, omni-directional CP antennas which are fed using various feed techniques are reported.

I. SINGLE FREQUENCY OMNIDIRECTIONAL CP ANTENNA DESIGN

In order to be integrated successfully into various wireless systems, omni-directional CP antennas need to be fed using various feedlines, such as CPW, microstrip and balanced lines. These feedline types are presented for both single and dual-band omni-directional antennas. In this work, the substrate used was Taconic RF-35 with a relative permittivity 3.47, a loss tangent 0.0018 and thickness of 1.57 mm.

A. CPW-Fed Omnidirectional CP Antenna Design

The antenna consists of a two layer substrate and three layers of metallisation: the two outer layers forming the patches, the middle layer forming the ground plane with CPW feed [14]. The patches are connected by a copper conducting strip as shown in Fig. 1 (a). The structure was simulated by using CST MWS. The parameters of the optimised antenna are as follows (mm): $l_a = 51.2$, $l_b = 49.6$, $W_s = 14$, $S = 55.8$, $\Delta_a = 2.3$, $\Delta_b = 3.1$, $\Delta_r = 19.5$, $\Delta_r = 2$, $l_t = 12$, $\Delta_s = 1$.

Fig.1 (b) shows that the measured and simulated S_{11} are in agreement. The measured S_{11} is < -10 dB from 1.579 GHz to 1.613 GHz has a 2.1% impedance bandwidth. RHCP is achieved (AR < -4 dB) in the band 1.593 - 1.601 GHz, providing an omnidirectional CP bandwidth of 7 MHz (0.4 %). The measured and simulated AR against angle at 1.596 GHz is shown in Fig. 2 (a).

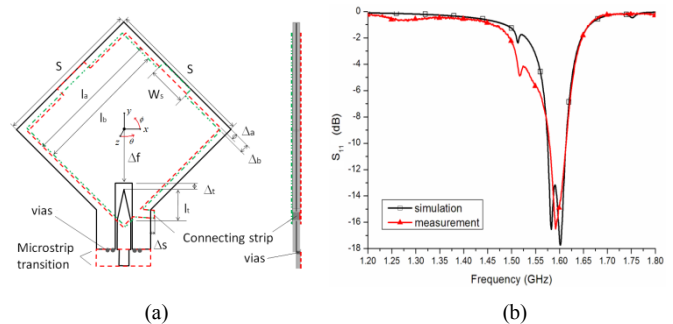
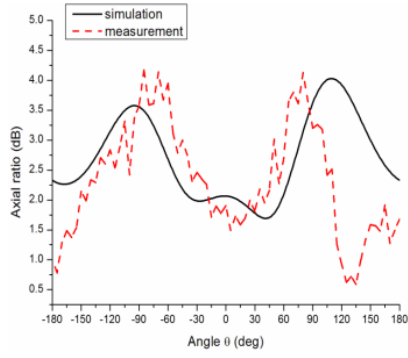
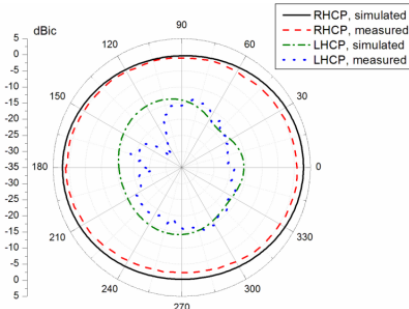


Fig.1(a) Configuration of the proposed CPW-fed omnidirectional CP antenna and (b) the simulated and measured results



(a)



(b)

Fig. 2(a) AR against angle and (b) patterns at 1.576 GHz

The AR is better than ~ 4 dB over the full 360° and generally better than 3.5 dB. The measured peak gain is 1.1 dBic, with a variation of ~ 3 dB in the $\Phi = 0^\circ$ plane. The cross-polar isolation is better than 13 dB as shown in Fig. 2 (b).

B. Microstrip-Fed Omnidirectional CP Antenna

Three different microstrip-fed antenna geometries are proposed in Figs 3-5 which are printed on two-layer substrates. An omni-directional circular patch arrangement is shown in Fig 3 where the top substrate size is $W_1 \times L_1$ and the lower one is $W_2 \times L_2$. Two stubs at 45° with respect to feedline are placed on both patches. A via of radius r_v is connected between the top and bottom patch located at (x, y) , where $(0, 0)$ corresponds to the centre of the patch. In order to isolate the via from the groundplane, a circular slot of radius r_s is appropriately located on the centre plane. By adjusting key parameters, such as stubs W_1 and L_1 and the via location, an omni-directional CP performance is realised. The parameter values are tabulated in Table 1.

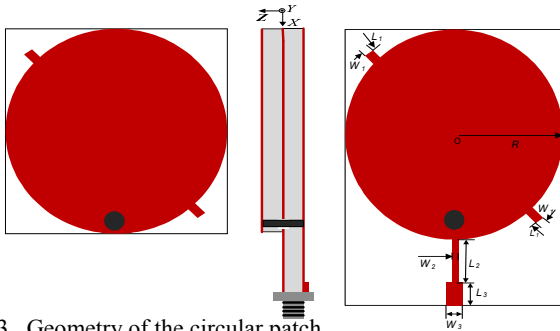


Fig. 3. Geometry of the circular patch

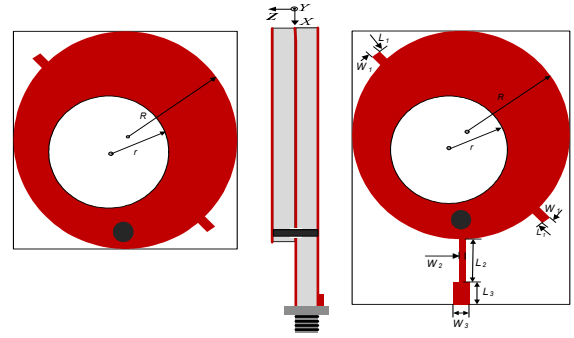


Fig.4. Geometry of the circular patch with slot circular

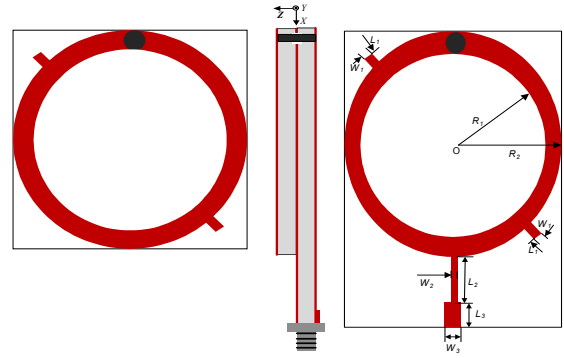


Fig. 5. Geometry of the annular ring patch

Table 1. Parameters of Microstrip-fed antennas

Antenna Types Parameters (mm)	Circular Patch	Circular Patch With circular Slot	Annular Ring Patch
R	28.7	25.0	-
r	-	10.7	-
R_1	-	-	20.55
R_2	-	-	15.85
W_1/L_1	2/3.5	2/3.5	2/3.5
W_2/L_2	2/10	0.6/12	0.6/12.46
W_3/L_3	3/7	3/2	3/2
Via location	(22,0)	(18,0)	(-17.8,0)
Origin of Slot		(2.7,2.7)	-
Substrate Size	62X62 70X62	55X50 65X50	42X42 65X42

Fig. 6 shows the simulated S_{11} for the three antennas. It is noted that the impedance bandwidth is increased as the dimension of antenna is increased. The three antennas provide an axial-ratio (AR) less than 3 dB in the YoZ plane at

1.575 GHz, as shown in Fig. 7. The annular-ring patch is the smallest geometry and is 44 % less than the circular patch antenna. Fig. 8 illustrates the radiation patterns at 1.575 GHz for the YoZ plane. As seen, RHCP pattern in the YoZ plane is omni-directional.

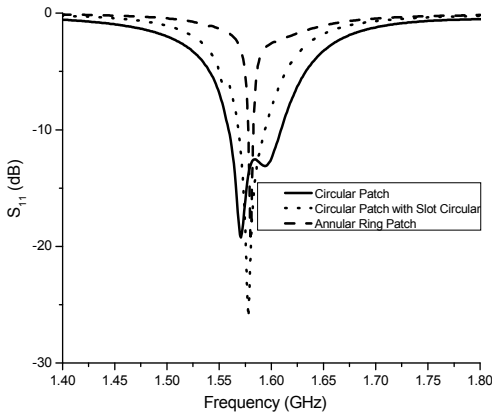


Fig.6 The simulated S_{11} for microstrip-fed omni-directional CP antennas

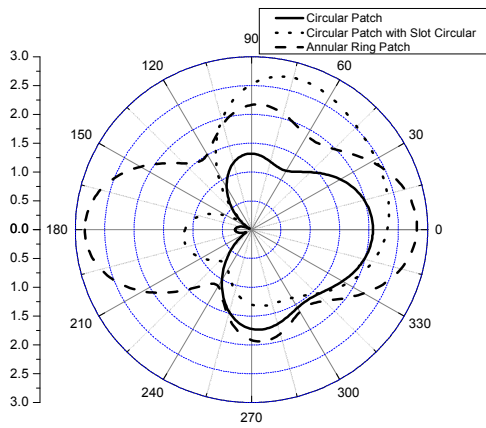


Fig. 7. Axial-ratio for the microstrip-fed patch antennas in the YoZ plane

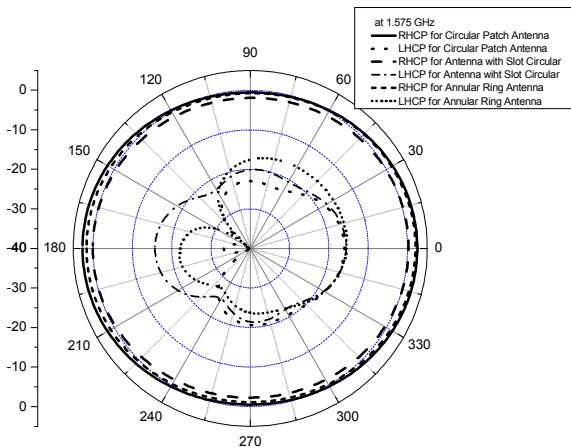


Fig. 8 Radiation patterns in the YoZ plane

C. Differentially-fed Omnidirectional CP Antenna

The proposed antenna again consists of a two layer substrate with three metallisation layers. The central ground plane measures $L_x \times L_y$ and the top and bottom layers have the same square patch structure with a 45 degree slot located centrally, which excites two nearly degenerate resonant modes and achieve CP performance. The configuration is shown in Fig. 9. The balanced parallel lines present a 300Ω input impedance. The substrate height is 3.175 mm.

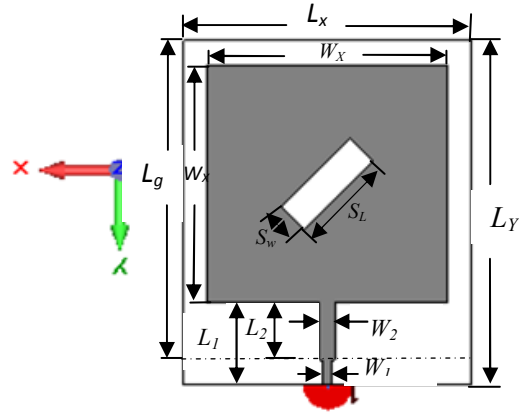


Fig. 9. The configurations of the proposed differentially-fed antenna $L_x=35.6$ mm, $L_y=42.8$ mm, $L_g=37.8$ mm, $S_l=12$ mm, $S_w=4$ mm, $W_x=29.6$ mm, $L_1=10.2$ mm, $L_2=7.2$ mm, $W_1=1$ mm, $W_2=2$ mm

Fig. 10 shows a simulated S_{11} bandwidth of 70 MHz from 2.374 GHz to 2.444 GHz. Fig. 11 displays the AR plotted against azimuthal angle in the YoZ plane. It is found that AR bandwidth is 25 MHz, from 2.418 GHz to 2.43 GHz. Fig 12 shows that the cross-polarization levels are approximately 20 dB over 360 degrees in the YoZ plane at 2.43 GHz.

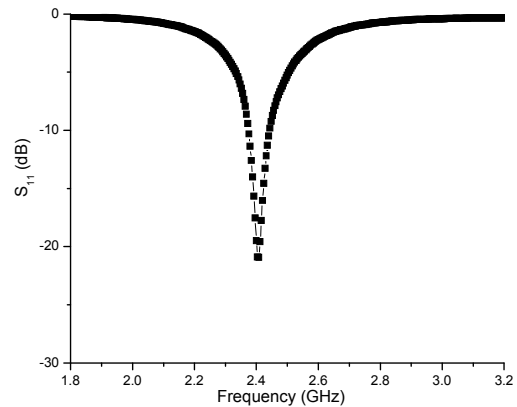


Fig.10. The simulated S_{11} for the proposed differentially-fed antenna

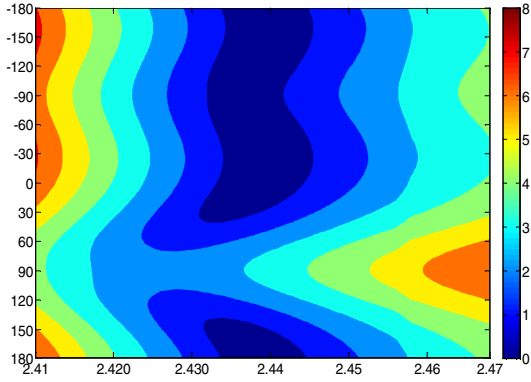


Fig. 11. The axial-ratio values plotted against angle

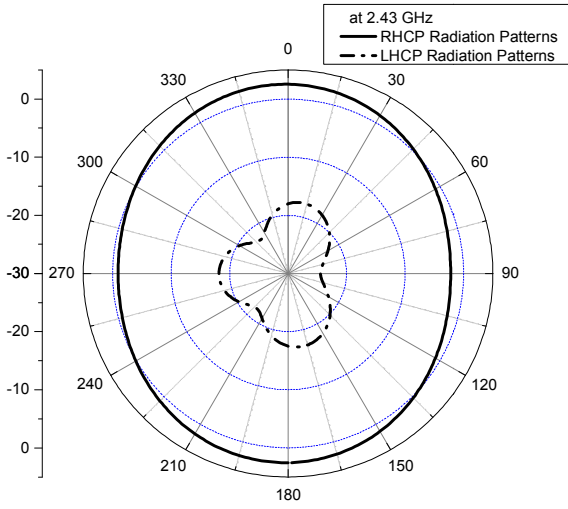


Fig. 12. Simulated radiation patterns for the proposed antenna

II. DUAL-FREQUENCY OMNIDIRECTIONAL CP ANTENNA DESIGN

Dual-frequency operation was also achieved for the antenna [15] with the addition of slots as shown in Figure 13. The radiators are electromagnetically-coupled to a 50Ω CPW located along the diagonal of the patches. The end of the line is triangularly tapered to increase its impedance. The parameter l_{CPW} was optimized to realize best matching. There are two slots cut in each patch, with a 1.5 pF lumped capacitor connected across the centre. The parameter values are shown in Figure 13. The S_{11} is shown in Figure 14. The simulated radiation efficiency is greater than 50% for both bands. Fig. 15 shows the AR as a function of both angle and frequency. The measured frequency ratio is 1.182. A clear omnidirectional and dual-band pattern can be seen in Fig. 16 (a) and (b). The RHCP gain ranges from -0.7 dBic to -4.5 dBic for the upper band and from -0.7 dBic to -4 dBic for the lower band.

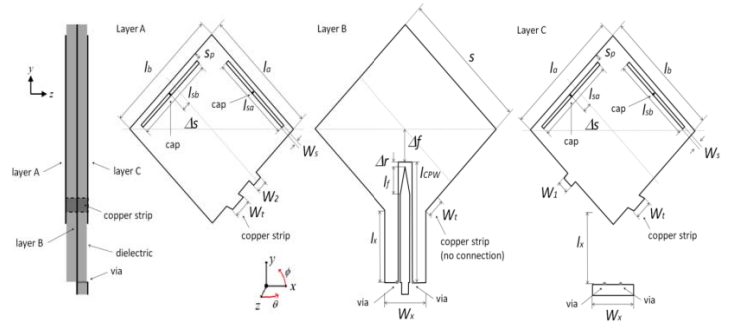


Fig. 13. The geometry of the proposed dual-frequency Omnidirectional CP antenna. Parameters: $l_a = 57 \text{ mm}$; $l_b = 57.1 \text{ mm}$; $s_p = 2.25 \text{ mm}$; $l_{sb} = 36.8 \text{ mm}$; $l_{sa} = 35.8 \text{ mm}$; $W_s = 1.5 \text{ mm}$; $\Delta_s = 3 \text{ mm}$; $W_t = 5.0 \text{ mm}$; $\Delta_f = 14.0 \text{ mm}$; $\Delta_r = 2.0 \text{ mm}$; $l_f = 12.0 \text{ mm}$; $l_{cpw} = 50.8 \text{ mm}$; $W_x = 20.0 \text{ mm}$; $l_x = 29.7 \text{ mm}$; $s = 64 \text{ mm}$.

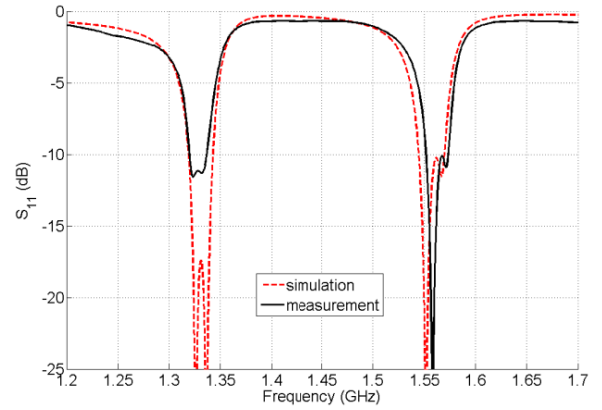


Fig. 14 Simulated and measured results for the dual-frequency omnidirectional CP antenna

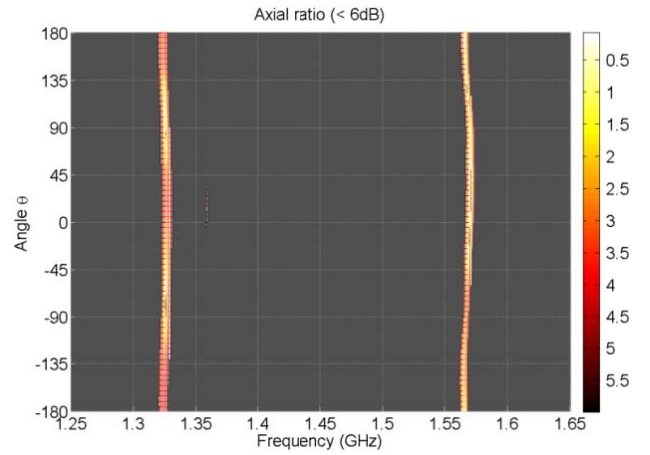


Fig. 15 The measured axial-ratio for the dual-frequency omnidirectional CP antenna

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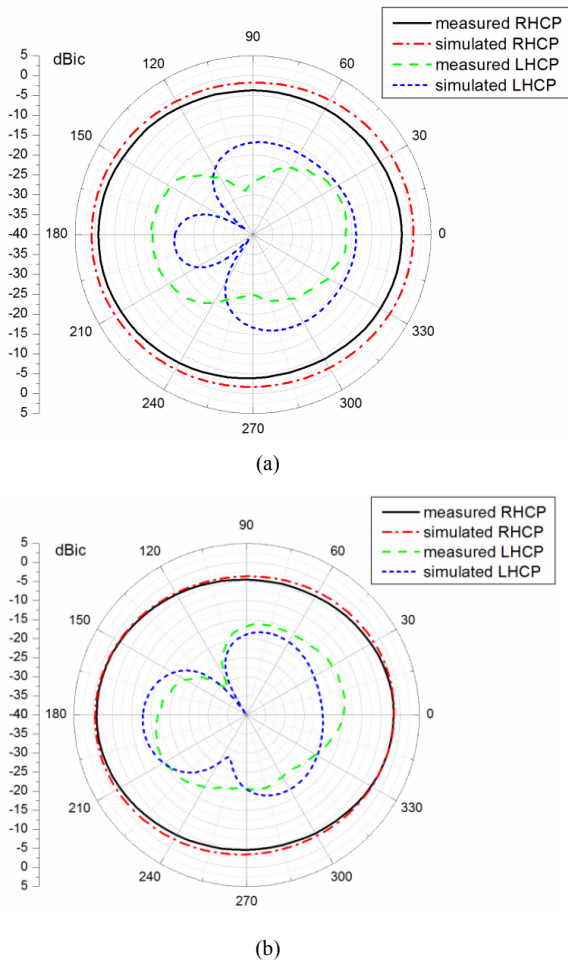


Fig. 16. (a) Simulated and measured radiation patterns for the low frequency and (b) high frequency.

III. CONCLUSIONS

Microstrip patch antennas are shown to provide omnidirectional CP using various feed techniques with AR below ~ 4 dB. For the microstrip-fed geometries, the annular-ring patch achieves the smallest size. A new technique to control the current distribution by bonding the back-to-back patches and the use of a diagonal CPW shallow inset feed was demonstrated to provide good matching and AR. A differentially-fed antenna is introduced with relatively broadband CP performance. Dual-frequency performance was achieved using lumped capacitor loaded slots which enable the generation of TM_{200} and TM_{020} modes.

ACKNOWLEDGMENT

This publication has emanated from research conducted with the financial support of Science Foundation Ireland under Grant Number 09/SIRG/I1644.

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